



A New Direction in Mixed-Signal

November 2012

SP7648

Low Reference High Efficiency Boost Regulator

Rev. 2.0.0

GENERAL DESCRIPTION

The SP7648 is an ultra-low quiescent current, high efficiency step-up DC-DC converter ideal for single cell Li-Ion or dual cell alkaline battery applications to drive various LEDs. The SP7648 combines low quiescent current and excellent light-load efficiency of PFM control. The SP7648 features synchronous rectification, a 0.3Ω charging switch, an anti-ringing inductor switch, under-voltage lockout and programmable inductor peak current. The device can be shut down by a 1nA active LOW shutdown pin. A very low 0.288V reference voltage is optimized for driving a constant current load.

APPLICATIONS

- LED Driver
- Camera Flash
- Handheld Portable Devices

FEATURES

- True Shutdown
- 700mA Output Current at 3.3V Input; 4.2V output
- 92% Efficiency from 2.7V_{IN} to 3.3V_{OUT}
- 2.7V to 4.5V Wide Input Voltage Range
- 5V Fixed or Adjustable Output
- 0.3Ω Switch
- Integrated Synchronous Rectifier: 0.3Ω
- Anti-Ringing Switch Technology
- Programmable Inductor Peak Current
- Logic Shutdown Control
- Low 0.8V or 0.288V Reference Voltage
- Small 10-pin DFN and MSOP Package

TYPICAL APPLICATION DIAGRAM

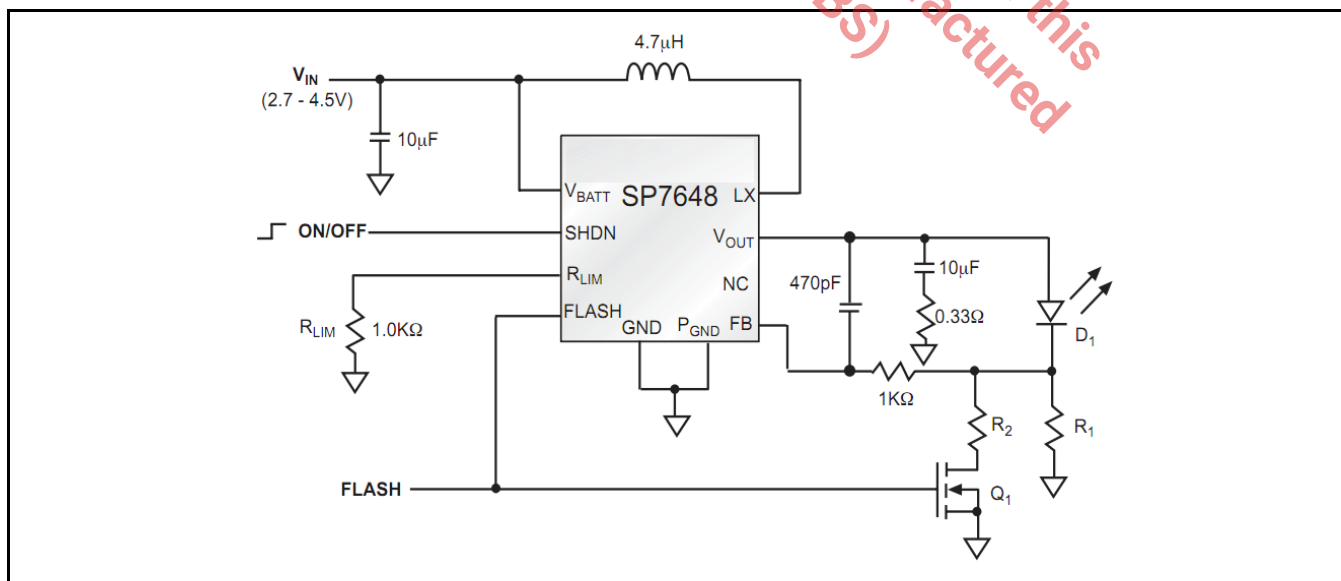


Fig. 1: SP7648 Application Diagram

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ABSOLUTE MAXIMUM RATINGS

These are stress ratings only and functional operation of the device at these ratings or any other above those indicated in the operation sections of the specifications below is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability.

LX, Vo, V_{BATT}, FLASHOUT, FB to GND pin..... -0.3 to 6.0V
 SHDN, FLASH..... -0.3V to V_{BATT}+1.0V
 VO, GND, LX Current 2A
 Reverse V_{BATT} Current..... 220mA
 Forward V_{BATT} Current..... 500mA
 Storage Temperature -65°C to 150°C
 ESD Rating (HBM - Human Body Model) 1.5kV

OPERATING RATINGS

Input Voltage Range V_{BATT}..... 2.7V to 4.5V
 Ambient Temperature Range -40°C to +85°C
 Thermal Resistance
 θ_{JA} (DFN-10)..... 40.5°C/W
 θ_{JA} (MSOP-10) 214.0°C/W

ELECTRICAL SPECIFICATIONS

Specifications with standard type are for an Operating Ambient Temperature of T_A = 27°C only; limits applying over the full Operating Ambient Temperature range are denoted by a "•". Minimum and Maximum limits are guaranteed through test, design, or statistical correlation. Typical values represent the most likely parametric norm at T_A = 27°C, and are provided for reference purposes only. Unless otherwise indicated, V_{BATT} = V_{SHDN} = 3.6V, V_{FB} = ZeroV, I_{LOAD} = 0mA, V_{OUT} = 5.0V, T_A = -40°C to 85°C.

Parameter	Min.	Typ.	Max.	Units	Conditions
Input Voltage Operating Range V _{BATT}	2.7	-	4.5	V	• After Startup
Output Voltage Range V _{OUT}	27	-	5.5	V	•
Under Voltage Lock-out UVLO	0.5	0.61	0.7	V	•
Output Voltage V _O	4.6	5.0	5.4	V	• Internal Feedback Divider
Shutdown Current into V _O , I _{SDO}	-	1	500	nA	• V _{SHDN} = ZeroV
Shutdown Current into V _{BATT} , I _{SDB}	-	250	750	nA	• V _{SHDN} = ZeroV, V _{BATT} = 2.7V
Efficiency		92	-	%	V _{BATT} = 2.7V, I _{OUT} = 200mA, R _{LIM} = 2kΩ
Inductor Peak Current Limit, I _{PK}	650	800	1000	mA	• R _{LIM} = 2kΩ, I _{PK} = 1600/R _{LIM}
		1600			• R _{LIM} = 1kΩ, I _{PK} = 1600/R _{LIM}
Output Current (Note 2)		800		mA	• V _{BATT} = 2.7V, R _{LIM} = 1kΩ
		400			• V _{BATT} = 2.7V, R _{LIM} = 2kΩ
		200			• V _{BATT} = 2.7V, R _{LIM} = 4kΩ
Minimum Off-Time Constant K _{OFF}	0.5	1.0	1.5	V*μs	• K _{OFF} ≤ T _{OFF} (V _{OUT} - V _{BATT})
Maximum On-Time Constant K _{ON}	2.0	3.5	5.0	V*μs	• K _{ON} ≥ T _{ON} (V _{BATT})
Enable Valid to Output Stable (Note 3)		300	500	μs	I _{LOAD} = 1mA
NMOS Switch Resistance		0.30	0.6	Ω	• I _{NMOS} = 100mA
PMOS Switch Resistance		0.30	0.6	Ω	• I _{PMOS} = 100mA
FB Set Voltage, V _{FB}	0.76	0.8	0.84	V	• External feedback Flash = 0
	0.266	0.288	0.310		• External feedback Flash = 1
FB Input Current	-	1	100	nA	• V _{FB} = 1.3V
SHDN Input Voltage V _{IL} (Note 1)			0.5	V	• V _{BATT} = 2.7V
SHDN Input Voltage V _{IH} (Note 1)	2.0				
SHDN Input Current		1	100	nA	•
LX Pin Leakage			3	μA	
FLASH Threshold V _{IL}			0.4	V	
FLASH Threshold V _{IH}	1.0				

Note 1: SHDN must transition faster than 1V/100mS for proper operation.

Note 2: Output Current I = (VIN/VOUT) x Efficiency x (Inductor Peak Current - Inductor Ripple Current)/2

Note 3: Guaranteed by Design.

BLOCK DIAGRAM

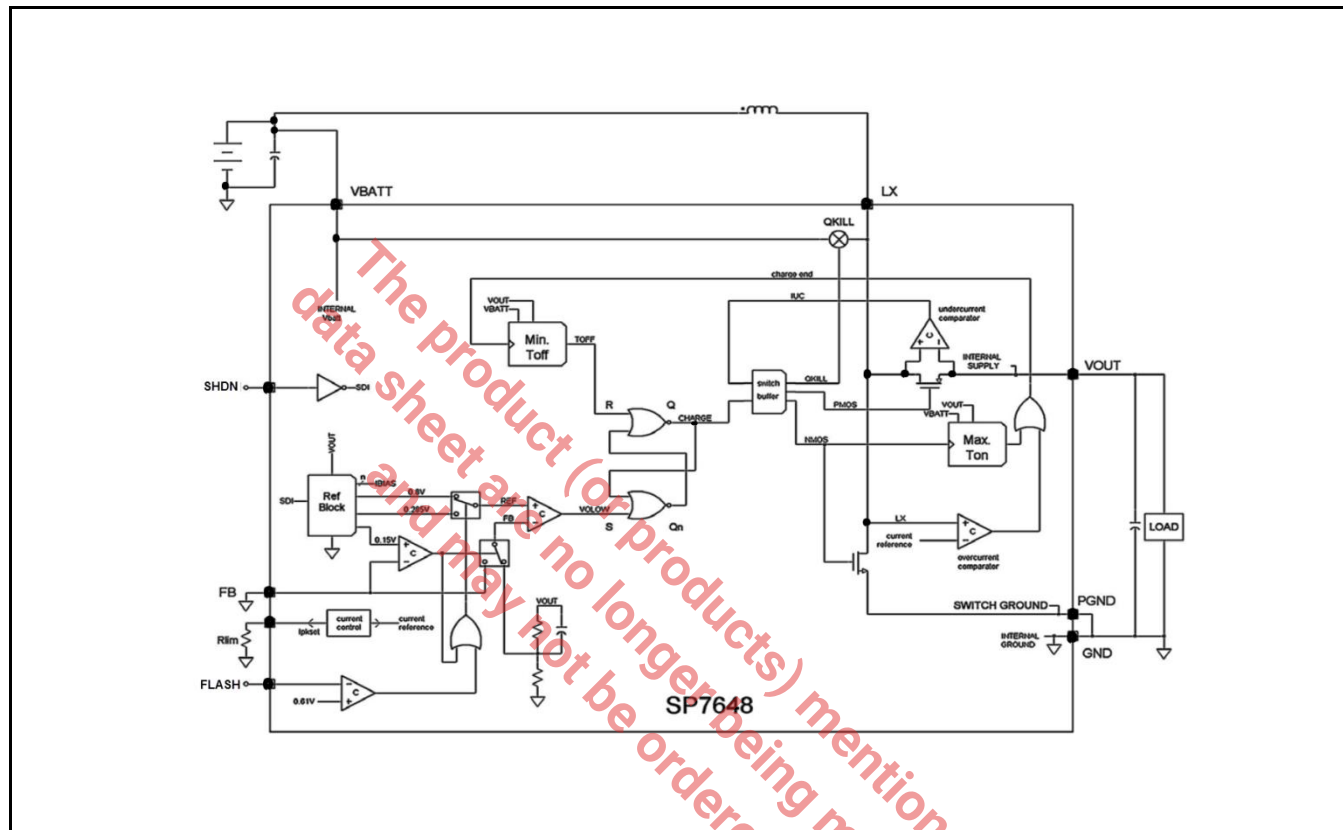


Fig. 2: SP7648 Block Diagram

PIN ASSIGNMENT

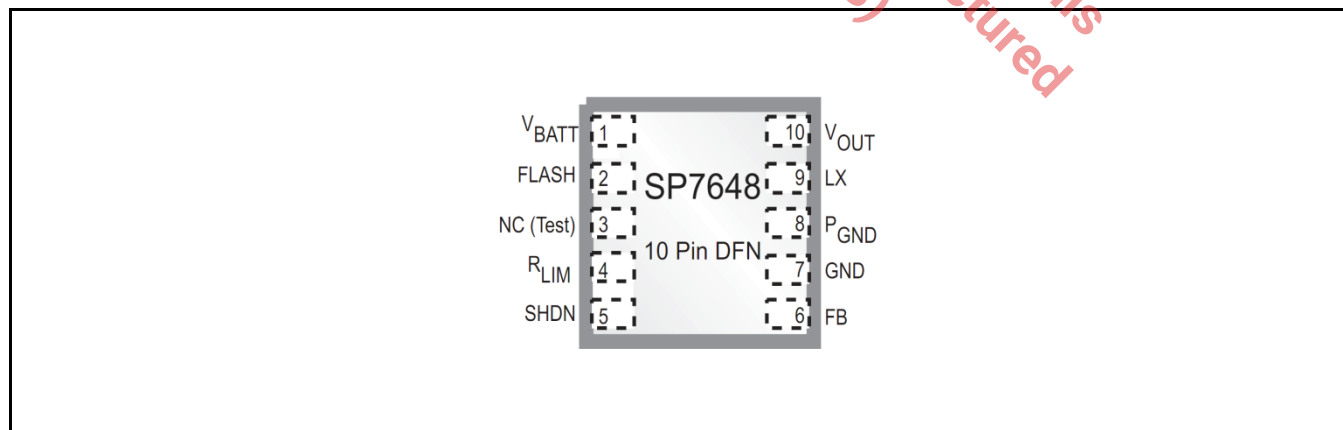


Fig. 3: SP7648 Pin Assignment

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PIN DESCRIPTION

Name	Pin Number	Description
V _{BATT}	1	Battery Voltage. The startup circuitry is powered by this pin. Battery Voltage is used to calculate switch off time: $T_{OFF} = K_{OFF} / (V_{OUT} - V_{BATT})$. When the battery voltage drops below 0.61V the SP7648 goes into an undervoltage lockout mode (UVLO), where the part is shutdown.
FLASH	2	Reference Control Input. Internal Reference defaults to 0.8V if FLASH = LOW and 0.288V if FLASH = HIGH.
NC (Test)	3	No connection. This pin is bonded out for test purposes only and must be left floating in all applications.
R _{LIM}	4	Current Limit Resistor. By connecting a resistor R _{LIM} from this pin to ground the inductor peak current is set by $I_{PEAK} = 1600 / R_{LIM}$. The range for R _{LIM} is 9kΩ (for 180mA) to 1KΩ (for 1.6A).
SHDN	5	Shutdown Not. Tie this pin high to V _{BATT} , for normal operation. Pull this pin to ground to disable all circuitry inside the chip.
FB	6	Feedback. Connect this pin to GND for fixed +5V operation. Connect this pin to a resistor voltage divider between V _{OUT} and GND for adjustable output operation.
GND	7	Ground. Connect to ground plane.
PGND	8	Power Ground. The inductor charging current flows out of this pin.
LX	9	Inductor Switching Node. Connect one terminal of the inductor to the positive terminal of the battery. Connect the second terminal of the inductor to this pin. The inductor charging current flows into LX, through the internal charging N-channel FET, and out the PGND pin.
V _{OUT}	10	Output Voltage. The inductor current flows out of this pin during switch off-time. It is also used as the internal regulator voltage supply. Connect this pin to the positive terminal of the output capacitor.
GND	Thermal Pad	Connect to ground signal.

ORDERING INFORMATION

Part Number	Ambient Temperature Range	Marking	Package	Packing Quantity	Note 1	Note 2
SP7648ER-L	-40°C ≤ T _A ≤ +85°C	SP76 48ER WWX	DFN-10	Bulk		
SP7648ER-L/TR				3K/Tape & Reel		
SP7648EU-L	-40°C ≤ T _A ≤ +85°C	7648 EXXX YWW	MSOP-10	Bulk		
SP7648EU-L/TR				2.5K/Tape & Reel		

"YY" = Year – "WW" = Work Week – "X" = Lot Number; when applicable.

TYPICAL PERFORMANCE CHARACTERISTICS

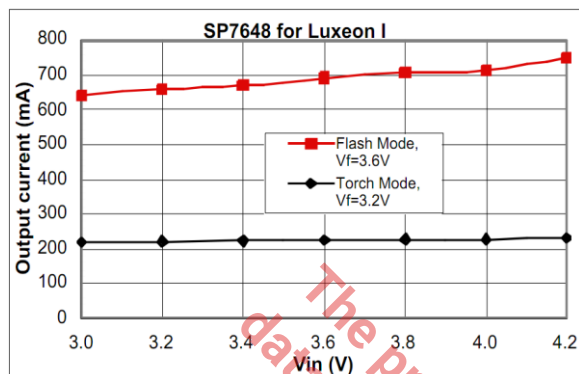


Fig. 4: Output Current versus Input Voltage

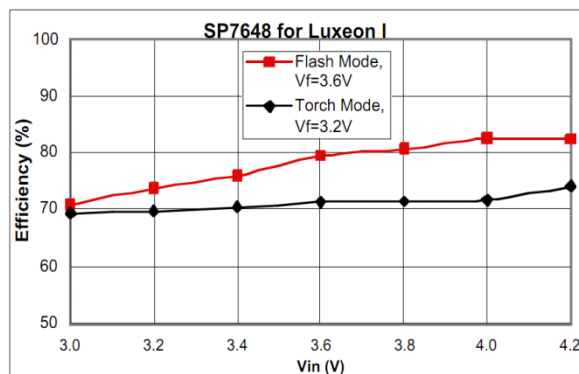


Fig. 5: Efficiency versus Input Voltage

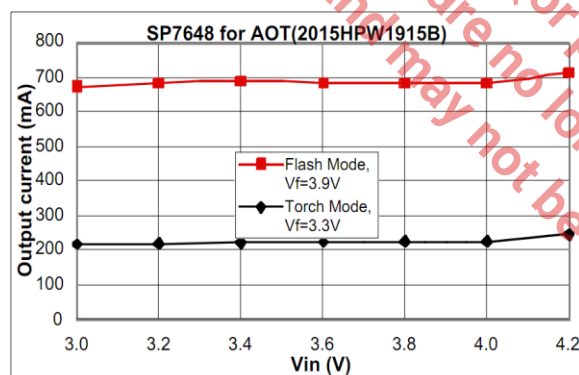


Fig. 6: Output Current versus Input Voltage

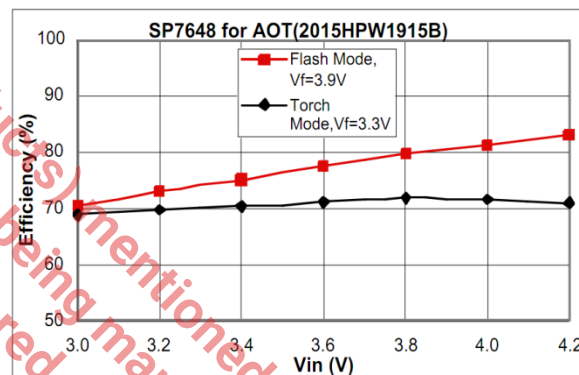


Fig. 7: Efficiency versus Input Voltage

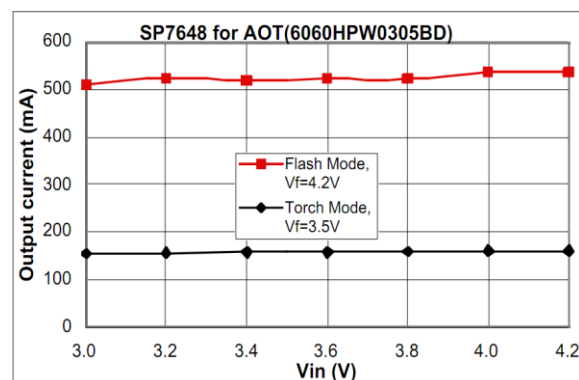


Fig. 8: Output Current versus Input Voltage

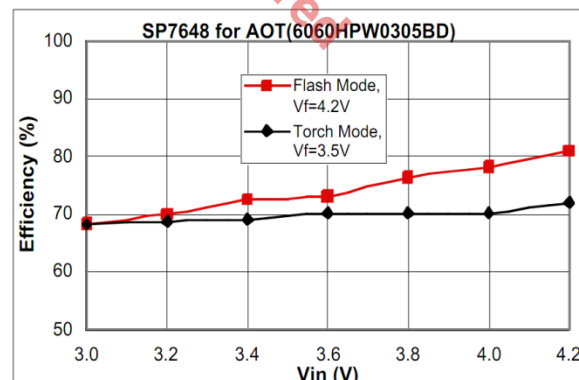


Fig. 9: Efficiency versus Input Voltage

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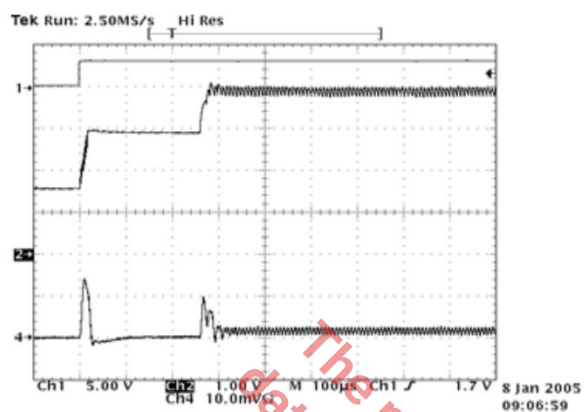


Fig. 10: Startup 200mA Torch
 $V_{IN} = 3.6V$, $V_{OUT} = 3.9V$, CH1=SHDN(5V/div),
CH2=VOUT(1V/div), CH3=Iin(1A/div)

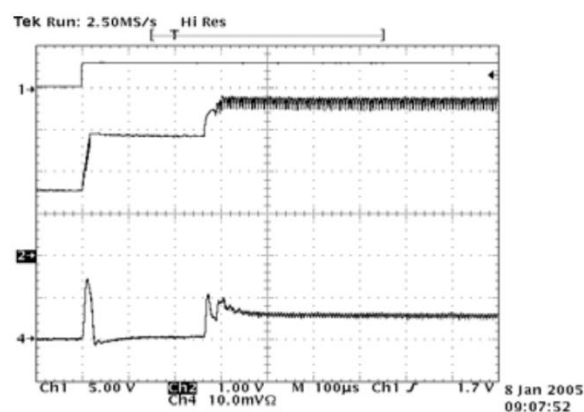


Fig. 11: Startup 700mA Flash
 $V_{IN} = 3.6V$, $V_{OUT} = 3.65V$, CH1=SHDN(5V/div),
CH2=VOUT(1V/div), CH3=Iin(1A/div)

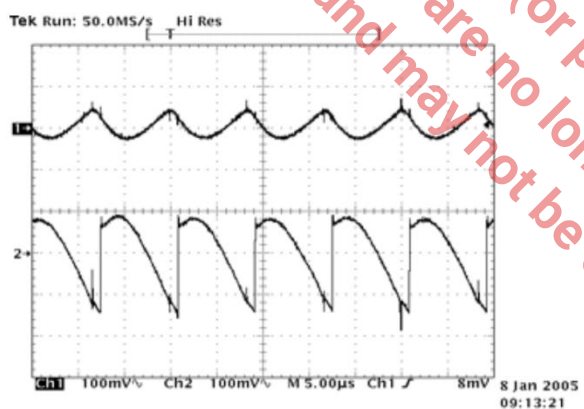


Fig. 12: Ripple 200mA Torch
 $V_{IN} = 3.6V$, $V_{OUT} = 3.9V$,
CH1=VIN(AC 100mV/div), CH2=VOUT(AC 100mV/div)

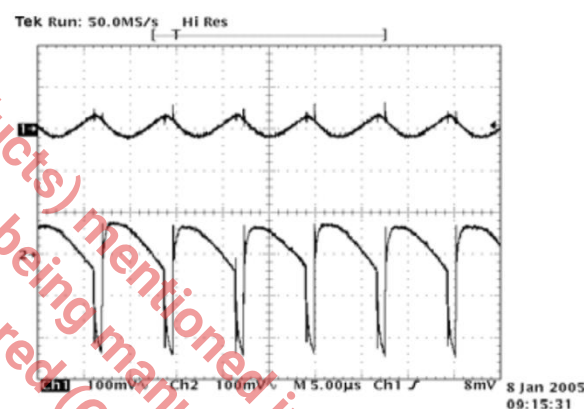


Fig. 13: Ripple 700mA Flash
 $V_{IN} = 3.6V$, $V_{OUT} = 3.65V$,
CH1=VIN(AC 100mV/div), CH2=VOUT(AC 100mV/div)

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THEORY OF OPERATION

DETAILED DESCRIPTION

The SP7648 is a step-up DC-DC converter with an input voltage operation range from 2.7V to 4.7V. In addition to the main 0.3Ω internal NMOSFET switch the SP7648 has an internal synchronous rectifier, thereby increasing efficiency and reducing the space and cost of an external diode. An internal inductive-damping switch significantly reduces inductive ringing for low noise-high efficiency operation. If the supply voltage drops below 0.61V the SP7648 goes into under voltage lockout, thus opening both internal switches. The inductor peak current is externally programmable to allow for a range of inductor values.

CIRCUIT LAYOUT

Printed circuit board layout is a critical part of a power supply design. Poor designs can result in excessive EMI on the feedback paths and on the ground planes with applications involving high switching frequencies and large peak currents. Excessive EMI can result in instability or regulation errors. All power components should be placed on the PC board as closely as possible with the traces kept short, direct, and wide (>50mils or 1.25mm). Extra copper on the PC board should be integrated into ground as a pseudo-ground plane. On a multilayer PC board, route the star ground using component-side copper fill, then connect it to the internal ground plane using vias. For the SP7648 devices, the inductor and input & output filter capacitors should be soldered with their ground pins as close together as possible in a star-ground configuration. The V_{OUT} pin must be bypassed directly to ground as close to the SP7648 devices as possible (within 0.2in or 5mm). The DC-DC converter and any digital circuitry should be placed on the opposite corner of the PC board as far away from sensitive RF and analog input stages. Noisy traces, such as from the LX pin, should be kept away from the voltage-feedback V_{FB} node and separated from it using grounded copper to minimize EMI. See the SP7648EB Evaluation Board Manual for PC Board Layout design details.

CONTROL SCHEME

A minimum off-time, current limited pulse frequency modulation (PFM) control scheme combines the high output power and efficiency of a pulse width modulation (PWM) device with the ultra low quiescent current of the traditional PFM. At low to moderate output loads the PFM control provides higher efficiency than traditional PWM converters are capable of delivering. At these loads the switching frequency is determined by a minimum off-time ($T_{OFF, MIN}$) and a maximum on-time ($T_{ON, MAX}$) where:

$$T_{OFF} \leq K_{OFF} / (V_{OUT} - V_{BATT})$$

$$T_{ON} \geq K_{ON} / V_{BATT}$$

$$K_{OFF} = 1.0V\mu s$$

$$K_{ON} = 3.5 V\mu s$$

At light loads (as shown in plot A in Figure 14) the charge cycle will last the maximum value for t_{ON} : For a 3V battery this would be as follows: $T_{ON} = K_{ON} / V_{BATT} = 3.5V\mu s / 3V = 1.17\mu s$. The current built up in the coil during the charge cycle gets fully discharged in the discontinuous conduction mode (DCM). When the current in the coil has reached zero, the synchronous rectifier switch is opened and the voltage across the coil (from V_{BATT} to LX) is shorted internally to eliminate inductive ringing. With increasing load (as shown in plot B in Figure 14) this inductor damping time becomes shorter, because the output will quickly drop below its regulation point due to heavier load. If the load current increases further, the SP7648 enters continuous conduction mode (CCM) where there is always current flowing in the inductor. The charge time remains at maximum T_{ON} as long as the inductor peak current limit is not reached as shown in plot C in Figure 14. The inductor peak current limit can be programmed by tying a resistor R_{LIM} from the R_{LIM} pin to ground where:

$$I_{PEAK} = 1600 / R_{LIM}$$

When the peak current limit is reached the charge time is short-cycled. In plot D of Figure 14, the switch current reaches the peak current limit during the charge period which ends the charge cycle and starts the discharge

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cycle. However, full load is not yet achieved because at the end of the minimum discharge time the output was still within regulation. Maximum load is reached when this discharge time has shrunk to the minimum allowed value T_{OFF} as shown in Plot E of Figure 14.

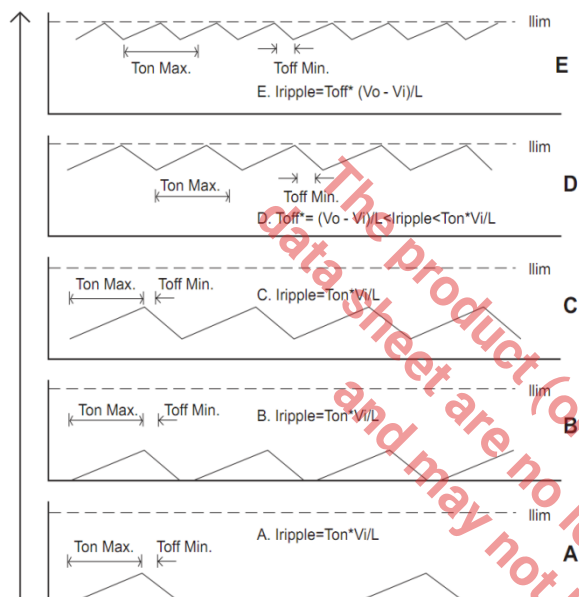


Fig. 14: Inductor Current versus Load

COMPONENT SELECTION

Selection of capacitors for SP7648 power supply circuits can be made through the use of the Component Selection Table. Capacitor equivalent series resistance (ESR) in the range of 0.2 to 0.3Ω is a requirement for obtaining sufficient output voltage ripple for the SP7648 to properly regulate under its load. For example, in the SP7648 application circuit a 10μF, 10V, X5R, surface mount ceramic output filter capacitor is used. Ceramic capacitors have an ESR too low to produce enough output ripple for the SP7648 to regulate the output; therefore, a 0.33Ω resistor is added in series with the 10μF capacitor at the V_{OUT} pin. Designers should select input and output capacitors with a rating exceeding the inductor current ripple, which is typically set by the inductor value and the K_{ON} value as given in the following relationship:

$$I_{L(RIPPLE)} = K_{ON}/L, \text{ where } K_{ON} = 3.5V \cdot \mu S$$

INDUCTORS - SURFACE MOUNT							
Inductance (uH)	Manufacturer/Part No.	Inductor Specification				Manufacturer Website	
		Series R ohms	Isat (A)	Size LxWxH (mm)	Inductor Type		
4.7	Würth Elektronik 744042004	0.070	1.70	4.8x4.8x1.8	Shielded Ferrite Core	www.we-online.de	
4.7	TDK SLF6025T-4R7M1R5-PF	0.031	1.50	6.0x6.0x2.5	Shielded Ferrite Core	www.tdk.com	
4.7	Coilcraft MSS6122-4R7MX	0.065	1.85	6.1x6.1x2.2	Shielded Ferrite Core	www.coilcraft.com	
6.8	Würth Elektronik 744042006	0.080	1.50	4.8x4.8x1.8	Shielded Ferrite Core	www.we-online.de	
6.8	TDK SLF6025T-6R8M1R3-PF	0.044	1.30	6.0x6.0x2.5	Shielded Ferrite Core	www.tdk.com	
6.8	Coilcraft MSS6122-6R8MX	0.100	1.45	6.1x6.1x2.2	Shielded Ferrite Core	www.coilcraft.com	
CAPACITORS - SURFACE MOUNT							
Capacitance (uF)	Manufacturer/Part No.	Capacitor Specification					Manufacturer Website
		ESR ohms (max)	Ripple Current 10C Rise (A)	Size LxWxH (mm)	Voltage (V)	Capacitor Type	
10	TDK C2012X5R0J106M	0.020	1.70	2.0x1.2x1.35	6.3	X5R Ceramic	www.tdk.com
10	Murata GRM21BR60J106KE19L	0.020	1.70	2.0x1.2x1.35	6.3	X5R Ceramic	www.murata.com
MOSFETS - SURFACE MOUNT							
MOSFET type	Manufacturer/Part No.	MOSFET Specification				Manufacturer Website	
		RDS(on) ohms	Id (A)	Package Size	MOSFET Specifications		
NMOS	Vishay Si1400DL	0.190	1.70	SC70-6	20V, Vgs = 2.5V, Pd = 0.5W	www.vishay.com	
NMOS	Fairchild FDN337N	0.070	2.20	SOT23-3	30V, Vgs = 2.5V, Pd = 0.5W	www.fairchildsemi.com	
RESISTORS - LOW VALUE SURFACE MOUNT							
Resistor Ref. Des.	Manufacturer/Part No.	Resistor Specification				Manufacturer Website	
		Resistance ohms	Pd (W)	Package Size	Resistor Specifications		
Rc, R2	Vishay CRCW0603R33F	0.33	0.10	0603	Thick Film 1% or 5%	www.vishay.com	
Rc, R2	Vishay CRCW0603R22F	0.22	0.10	0603	Thick Film 1% or 5%	www.vishay.com	

Note: Components highlighted in **bold** are those used on the SP7648EB Evaluation Board.

Table 1: Component Selection

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For the example, a 10 μ H inductor would have an inductor current ripple of 350mA, while a 4.7 μ H inductor would have an inductor current ripple value of 740mA. Do not allow tantalum capacitors to exceed their ripple-current ratings. An input filter capacitor can reduce peak currents drawn from the battery and improve efficiency. For most applications, use the same capacitor for the input and output. Low-ESR tantalum capacitors are acceptable provided they meet the ESR requirement of 0.2 Ω to 0.3 Ω . In selecting an inductor, the saturation current specified for the inductor needs to be greater than the SP7648 peak current to avoid saturating the inductor, which would result in a loss of efficiency and could damage the inductor. The SP7648 evaluation board uses a Würth 4.7 μ H inductor with an I_{SAT} value of 1.7A and a DCR of 0.065 Ω , which handles the I_{PEAK} of 1.6A of the SP7648 and will deliver high efficiencies. Other inductors could be selected provided their I_{SAT} is greater than the I_{PEAK} of the SP7648.

V_{OUT} PROGRAMMING

The SP7648 can be programmed as either a voltage source or a current source. To program the SP7648 as voltage source, the SP7648 requires 2 feedback resistors R1 & R2 to control the output voltage. To set V_{OUT} in the voltage mode, use the equation:

$$R1 = [(V_{OUT}/0.8)-1] \times R2 \text{ where flash} < 0.4V,$$

$$R1 = [(V_{OUT}/0.288)-1] \times R2 \text{ where flash} > 1.0V.$$

USING THE R_{LIM} FUNCTION

The peak inductor current, I_{PEAK} , is programmed externally by the R_{LIM} resistor connected between the R_{LIM} pin and GND. The peak inductor current is defined by:

$$I_{PEAK} = 1600/R_{LIM}$$

The saturation current specified for the inductor needs to be greater than the peak current to avoid saturating the inductor, which would result in a loss in efficiency and could damage the inductor. The SP7648 evaluation board uses a R_{LIM} value of 1K Ω for an I_{PEAK} = 1.6A to allow the circuit to deliver up to 700mA for V_{IN} = 3.3V and V_{OUT} = 4.2V. Other

values could be selected using the above relationships.

USING THE FLASH CONTROL PIN

The SP7648 will regulate the output by the equations above depending on the state of the FLASH pin. When the FLASH pin is low (<0.4V), the internal reference voltage is defined as 0.8V. When the FLASH pin is high (>0.4V), the internal reference voltage is defined as 0.288V. This allows the use of smaller values for the sense resistor for current regulation mode. This improves efficiency and reduces the physical size of the sense resistor. An external MOSFET switch can be used to change the sense resistor when changing to the Flash Mode.

HIGH BRIGHTNESS WHITE LED

For the high brightness LumiLED white LED application, the SP7648 is generally programmed as a current source. The bias resistors R1 and R2 are used to set the operating current of the white LED with the equation:

$$R = V_{FB}/I_F$$

where V_{FB} is 0.8V in torch mode and 0.288V in flash mode, I_F is the operating current of the LED. To set the operating current to be about 200mA in torch mode, the flash pin is forced low, R2 is selected as 0.8V/ 0.2 = 4 Ω , as shown in the typical application circuit. To set the operating current to 700mA in flash mode, the flash pin is forced high, R is selected as 0.288V/0.41 Ω = 700mA. In reality R in Flash includes the series MOSFET $R_{DS(on)}$ and the parallel combination of $R_2 = 4\Omega$ shown by the formula:

$$R \text{ in Flash} = (R_1 \times (R_2 + Q_1 R_{DS(on)})) / (R_1 + R_2 + Q_1 R_{DS(on)})$$

If the SP7648 is powered up before the LED is plugged in, the circuit will bring the feed-back pin to ZeroV and the SP7648 has a feature to set the output voltage to be 5V. Once the LED is plugged in, the feedback pin will go up to 0.8V in torch mode or 0.288V in flash mode and begin to regulate. The output voltage will go from 5V to $V_F + V_{FB}$, where V_F is the forward voltage of the LED. When the LED is open, the

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feedback pin voltage will go to ZeroV and the output voltage will go to 5V which will protect the part from overvoltage at the output.

One approach to control LED brightness is to apply a PWM signal to the SHDN input of the SP7648. In this case, the output current will be equal to the product of $V_{REF}/R1$ and the average duty cycle at the SHDN pin. An

optional 10K Ω potentiometer may also be used for dimming the LED current by varying the potentiometer between low brightness and full brightness.

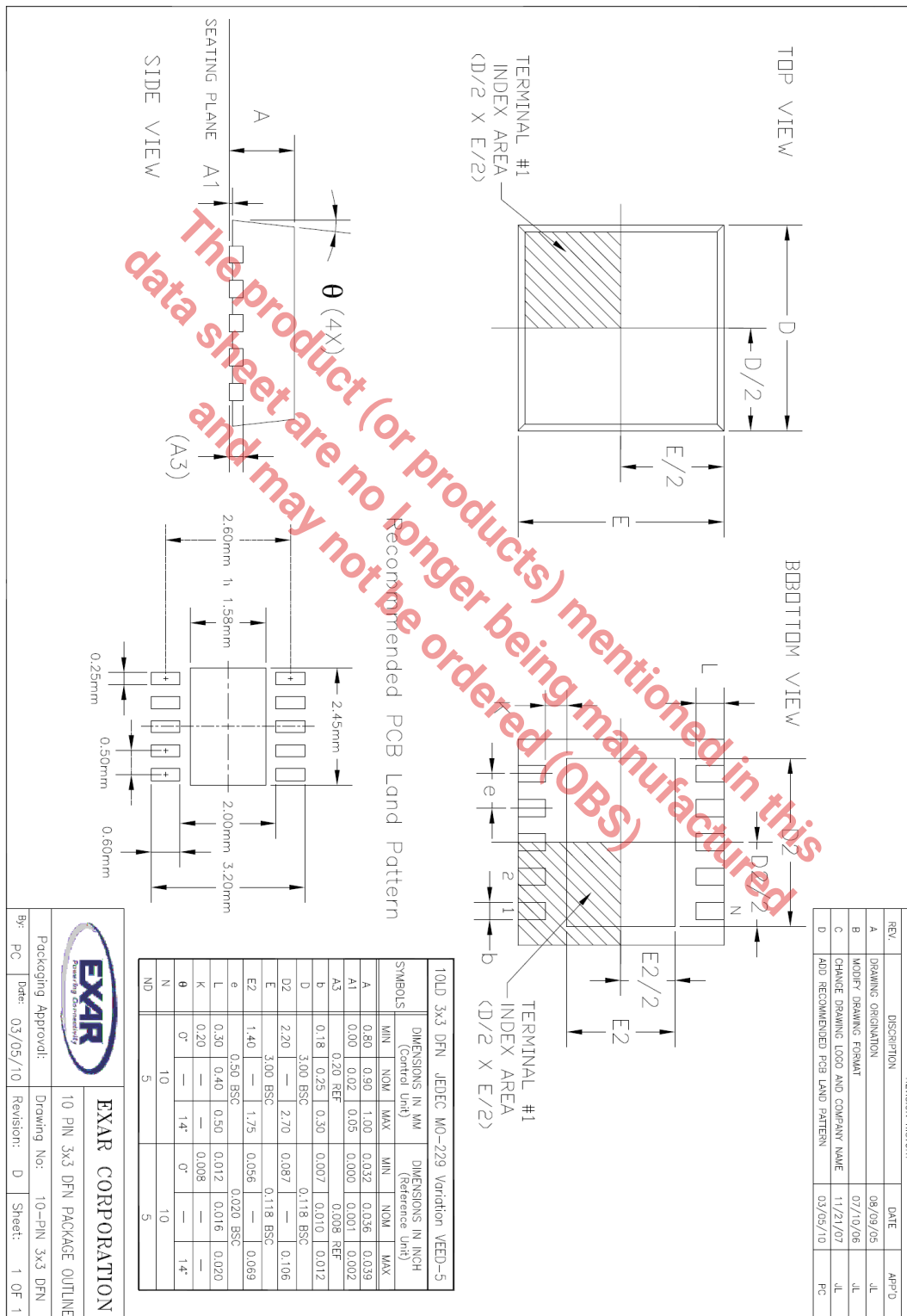
If the FB pin is pulled below 150mV the output will default to 5V defined by an internal resistor divider.

The product (or products) mentioned in this data sheet are no longer being manufactured and may not be ordered (OBS)

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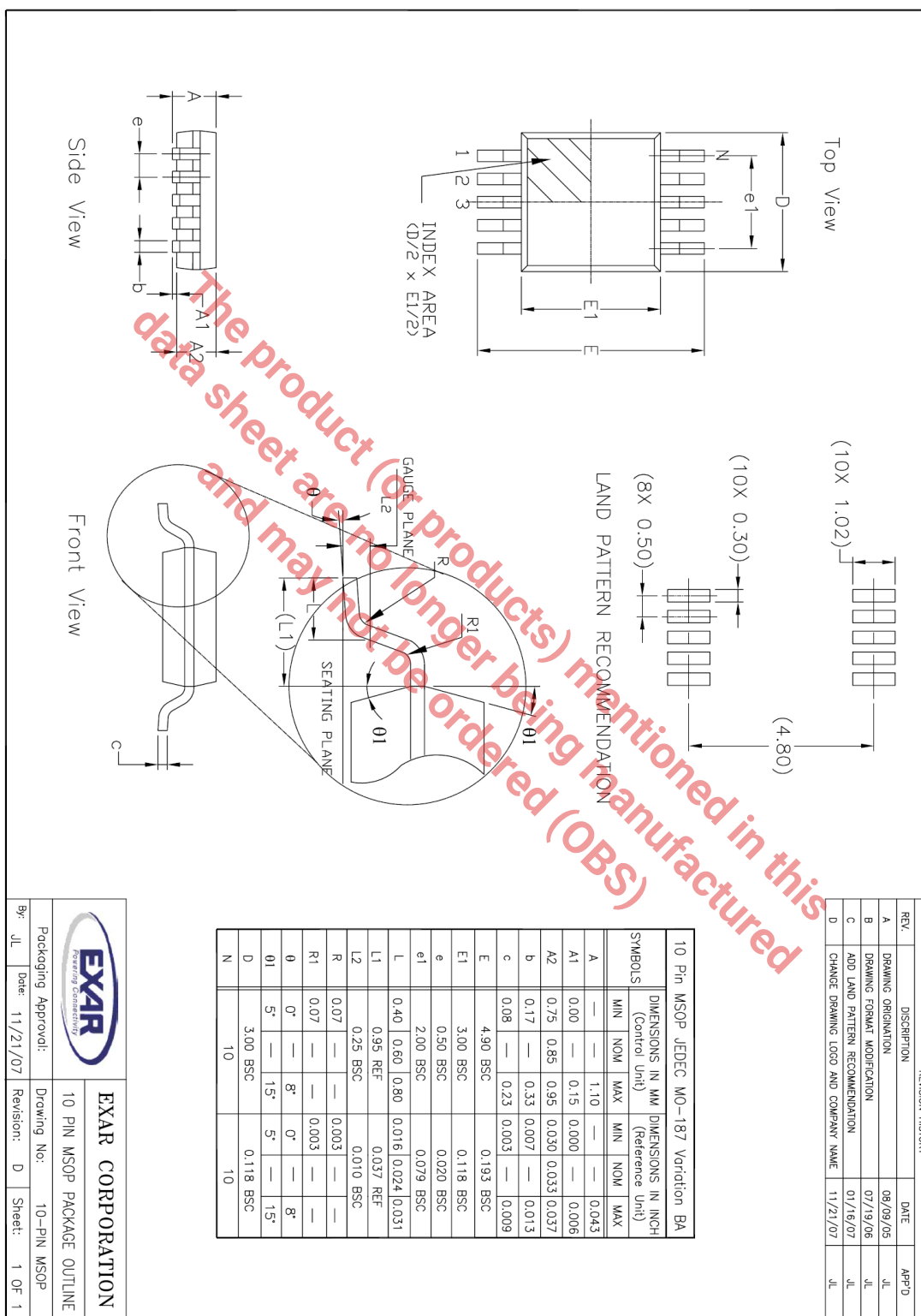
PACKAGE SPECIFICATION

10-PIN DFN



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10-PIN MSOP



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REVISION HISTORY

Revision	Date	Description
2.0.0	11/19/2012	Reformat of datasheet

FOR FURTHER ASSISTANCE

Email: customersupport@exar.com

Exar Technical Documentation: <http://www.exar.com/TechDoc/default.aspx?>



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